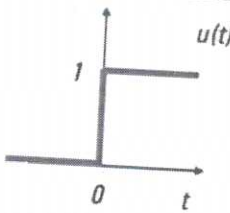


SCORING INDICATORS

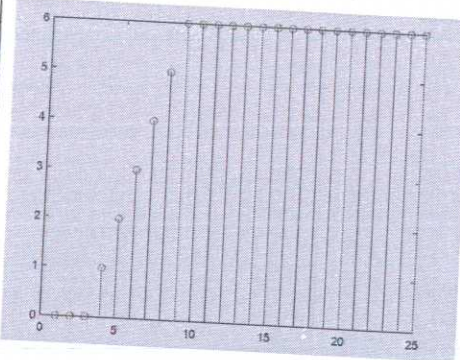
COURSE NAME : SIGNALS AND SYSTEMS (SET2)  
 COURSE CODE : TED (21) 5201

QID: 2109230106

Q No	Scoring Indicators	Split score	Sub Total	Total score
<b>PART A</b>				
I. 1	The signal whose energy is finite and average power is zero	1	1	9
I. 2	 <p>The time axis should be marked. The magnitude should be 1. Step at <math>t = 0</math> should be shown.</p>	1	1	
I. 3	<p>The signals that follow the even symmetry with respect to the y-axis(magnitude) are called even signals.</p> <p>The signals which follow the relation <math>x(t) = x(-t)</math> for all <math>t</math> or <math>x(n) = x(-n)</math> for all <math>n</math> are called even signals</p> <p><b>One of these two definitions</b></p>	1	1	
I. 4	$x(n) = \begin{cases} 0, & \text{for } n < 0 \\ 1, & \text{for } 0 \leq n \leq 5 \\ 0, & \text{for } n > 5 \end{cases}$ <p>Or <math>x(n) = \sin(\omega n)</math> for all <math>n</math></p> <p>In any one of the forms <math>x(n)</math> should be represented as a function of <math>n</math></p>	1	1	
I. 5	The system whose output at any time depends only on present and past values of input but not on future inputs	1	1	
I. 6	<p>Any one equation in the form</p> $y(n) = a.x(n) + b.x(n-1) + c.y(n-1) + \dots$ <p><math>a, b, c,</math> and etc. can be numbers also.</p>	1	1	
I. 7	DTFT	1	1	
I. 8	$2.f_{\max}$	1	1	
I. 9	$e^{-at}u(t),$ $e^{-at}$ can also be give full mark	1	1	
<b>PART B</b>				
II. 1	<p>Causal signals: the signals which does not exists or has zero magnitude for negative values of time ( <math>t</math> or <math>n</math>) are called causal signals. A continuous-time signal is said to be causal if <math>x(t) = 0</math> for <math>t &lt; 0</math>.</p> <p>Anti-causal signals: the signals that do not exist or have zero</p>	1 x 3	3	24

	<p>magnitude for positive values of time ( <math>t</math> or <math>n</math>) are called anti-causal signals. A continuous-time signal is said to be causal if <math>x(t) = 0</math> for <math>t &gt; 0</math>.</p> <p>Non-causal signals: all the signals which are not causal are called non-causal signals. Non-causal signals include anti-causal signals and the signals that exist for both positive and negative values of time.</p>				
II. 2	<p>Time scaling is either compression or expansion. (1 marks)</p> <p>Explanation 2 marks</p> <p>Replacing 't' with 'At' in a signal will cause expansion of the signal if 'A' is a positive constant with a magnitude less than 1.</p> <p>If 'A' is a positive constant with a magnitude greater than 1, it will cause time compression of the signal.</p> <p>If A is negative in addition to expansion or compression the signal will be time reversed.</p> <p><b>(full marks may be given even if time reversal is not mentioned. Diagrams are not necessary)</b></p>		1+2	3	
II. 3	<p><b>Plot of any 3 continuous time signals from the following,</b> with time axis and amplitude clearly labeled.</p> <ol style="list-style-type: none"> <li>Unit step signal <math>[u(t)]</math></li> <li>Unit impulse signal <math>[\delta(t)]</math></li> <li>Unit ramp signal <math>[r(t)]</math></li> <li>Unit parabolic signal <math>[p(t)]</math></li> <li>Sinusoidal signal</li> <li>Signum function</li> <li>Real exponential signal</li> <li>Rectangular pulse</li> <li>Triangular pulse</li> </ol>		1x 1	3	
II. 4	Periodic Signal	Aperiodic Signal	1	3	
	A definite pattern of repetition at regular interval	No pattern of repetition			
	A finite value of fundamental period. $x(t+T) = x(t)$ or $x(n+N) = x(n)$	Ideally period is infinite	1		
	eg:- $\sin(\omega t)$ sinusoids	eg:- $u(t)$ rectangular pulse, triangular pulse, and etc.	1		
II. 5	CT System	DT System	1 x 3	3	
	Input and output are CT signals	Input and output are DT signals			

	Described by a differential equation	Described by a difference equation			
	Impulse response is denoted as $h(t)$	Impulse response is denoted as $h(n)$			
	No such parameter, any continuous time signal can be processed	Sampling frequency of input is an important parameter to process			
	eg:- Audio or video Amplifiers	eg:- microprocessors, microcontrollers, DSP processors			
II. 6	<p>The response of a system to impulse is called impulse response. It is the output of the system when the input is the unit impulse signal</p> <p>The impulse response of a causal system is a causal signal. That means the impulse response has non-zero amplitude only for the positive values of time [ or <math>h(t) = 0</math> for <math>t &lt; 0</math>)]</p> <p>For a stable system the impulse response is a bounded signal. That means the impulse response is absolutely integrable.</p>		1+1+1	3	
II. 7	<p>For a system to be invertible, there should be a system such that which recovers the input from the output of the original system.</p> <p>a) not invertible, since the original input cannot be recovered after squaring the signal. The sign of the signal cannot be recovered</p> <p>b) invertible. The original system only introduces a time delay. The inverse system can be implemented as a system introducing a equal time advancement; ie., <math>y(n) = x(n+2)</math></p>		$1\frac{1}{2} \times 2$	3	
II. 8	<p><b>EXISTENCE OF FOURIER TRANSFORM</b></p> <p>There are a set of sufficient <u>conditions</u> called Dirichlet <u>conditions</u> under which a continuous-time signal <math>x(t)</math> is guaranteed to have a Fourier transform:</p> <ol style="list-style-type: none"> <li><math>x(t)</math> is absolutely integrable: <math>\int_{-\infty}^{\infty}  x(t)  dt &lt; \infty</math></li> <li><math>x(t)</math> has a finite number of maxima and minima in any finite interval.</li> <li><math>x(t)</math> has a finite number of discontinuities in any finite interval, and each of these discontinuities is finite.</li> </ol>		1x3	3	
II.9	<p>Statement 1 marks and explanation 2 mark</p> <p>The final value theorem states that if <math>X(s)</math> is the Laplace transform of <math>x(t)</math> then</p> $x(\alpha) = \lim_{s \rightarrow 0} sX(s)$ <p>Final value theorem is applied to find <math>x(\alpha)</math> when <math>x(t)</math> cannot be evaluated at <math>t = \alpha</math></p> <p>Final value theorem is applied when all the poles of <math>sX(s)</math> lies</p>		1+2	3	

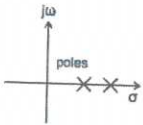
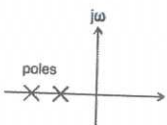
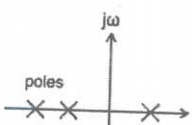
	on the left half of s-plane			
II.10	<p>Any 3 points</p> <p><b>PROPERTIES OF ROC OF LAPLACE TRANSFORM</b></p> <ul style="list-style-type: none"> <li>❖ ROC contains strip lines parallel to <math>j\omega</math> axis in s-plane.</li> <li>❖ If <math>x(t)</math> is absolutely integral and it is of finite duration, then ROC is entire s-plane</li> <li>❖ If <math>x(t)</math> is a right sided sequence then ROC: <math>\text{Re}\{s\} &gt; \sigma_0</math>.</li> <li>❖ If <math>x(t)</math> is a left sided sequence then ROC: <math>\text{Re}\{s\} &lt; \sigma_0</math>.</li> <li>❖ If <math>x(t)</math> is a two-sided sequence then ROC is the combination of two regions.</li> </ul>	1x3		
<b>PART C</b>				
III.	<p>a) Continuous-time signals and discrete-time signals</p> <p>b) Even signals and odd signals</p> <p>c) Periodic signals and aperiodic signals</p> <p>d) Energy signals and power signals</p> <p>e) Real signals and complex(imaginary) signals</p> <p>f) Deterministic signals and non-deterministic signals</p> <p>List 2 marks</p> <p>Explanation 5 marks</p>	2+5	7	<b>42</b>
IV	<p><math>x(n)</math></p>  <p>The signal has zero magnitude for <math>t &lt; 0</math>, therefore the signal is causal</p> <p>Drawing the signal 5 marks</p> <p>Checking causality 2 marks</p>	5+2	7	
V	<p>For checking linearity from a difference equation, linear combination of output is compared with the output when the input is a linear combination.</p> <p>a) let <math>y_1(t) = T[x_1(t)] = x_1^2(t)</math> and <math>y_2(t) = T[x_2(t)] = x_2^2(t)</math></p> <p>Then the linear combination of outputs, is</p> <p><math>A.y_1(t) + B.y_2(t) = A.x_1^2(t) + B.x_2^2(t) \dots \dots \dots (1)</math></p> <p>Now the output when input is a linear combination of <math>x_1(t)</math> and <math>x_2(t)</math>, is</p>	3+4	7	3 marks

	<p> <math>T[A \cdot x_1(t) + B \cdot x_2(t)] = (A \cdot x_1(t) + B \cdot x_2(t))^2</math>  <math>A^2 \cdot x_1^2(t) + B^2 \cdot x_2^2(t) + A \cdot B \cdot x_1(t) \cdot x_2(t) \dots \dots \dots (2)</math>            Since eqn.(1) <math>\neq</math> eqn.(2) the system is not linear            b) let <math>y_1(t) = T[x_1(t)]</math> and <math>y_2(t) = T[x_2(t)]</math>            when <math>y(t)</math> in the difference equation is replaced with a linear combination of outputs, the LHS becomes  <math display="block">\frac{d}{dt}(A \cdot y_1(t) + B \cdot y_2(t)) + A \cdot y_1(t) + B \cdot y_2(t)</math> <math display="block">= \frac{d}{dt}(A \cdot y_1(t)) + A \cdot y_1(t) + \frac{d}{dt}(B \cdot y_2(t)) + B \cdot y_2(t)</math> <math display="block">= A \cdot \left( \frac{dy_1(t)}{dt} + y_1(t) \right) + B \cdot \left( \frac{dy_2(t)}{dt} + y_2(t) \right)</math> <math display="block">= A \cdot \left( x_1(t) \frac{dx_1(t)}{dt} \right) + B \cdot \left( x_2(t) \frac{dx_2(t)}{dt} \right) \dots \dots \dots (3)</math>            when <math>x(t)</math> in the difference equation is replaced with a linear combination of inputs, the RHS becomes  <math display="block">(A \cdot x_1(t) + B \cdot x_2(t)) \cdot \frac{d}{dt}(A \cdot x_1(t) + B \cdot x_2(t))</math> <math display="block">= A^2 \cdot \left( x_1(t) \frac{dx_1(t)}{dt} \right) + B^2 \cdot \left( x_2(t) \frac{dx_2(t)}{dt} \right) +</math> <math display="block">A \cdot B \cdot \left( x_1(t) \frac{dx_2(t)}{dt} \right) + A \cdot B \cdot \left( x_2(t) \frac{dx_1(t)}{dt} \right) \dots \dots \dots (4)</math>            Since eqn. (3) <math>\neq</math> eqn. (4) the system is not linear         </p>	4 marks		
VI	<p>           Any 4 properties of            1. Linearity            2. Time invariance            3. Causality            4. Possessing Memory (Static or Dynamic)            5. Stability            6. Invertibility            Listing 1 marks            Description 1½ marks each         </p>	1+1½ x 4	7	
VII	<p>           Statement 2 marks            5 marks for diagram and explanation            Sampling theorem: A bandlimited continuous time signal with maximum frequency <math>f_m</math> can be represented using its samples and can be recovered back when sampling frequency <math>f_s</math> is greater than or equal to the twice the highest frequency component, <math>f_m</math>.            i.e., <math>f_s \geq 2 \cdot f_m</math> </p>	2+5	7	



	<p><b>Frequency Shifting</b>            If <math>x(t) \xrightarrow{FS} c_n</math>            Then according to frequency shifting property,  <math>e^{jm\omega_0} x(t) \xrightarrow{FS} C_{(n-m)}</math></p> <p><b>Differentiation in Time</b>            According to this property, if <math>x(t) \xrightarrow{FS} c_n</math>            Then <math>\frac{d}{dt} x(t) \xrightarrow{FS} (jn\omega_0) c_n</math></p>	2		
	<p><b>Differentiation in Time</b>            According to this property, if <math>x(t) \xrightarrow{FS} c_n</math>            Then <math>\frac{d}{dt} x(t) \xrightarrow{FS} (jn\omega_0) c_n</math></p>	2		
X	<p>a) <math>e^{-at} u(t)</math>  <math display="block">X(j\omega) = \int_{-\alpha}^{\alpha} e^{-at} u(t) e^{-j\omega t} dt = \int_0^{\alpha} e^{-at} e^{-j\omega t} dt</math> <math display="block">= \int_0^{\alpha} e^{-(a+j\omega)t} dt = \left[ \frac{e^{-(a+j\omega)t}}{-(a+j\omega)} \right]_0^{\alpha} = \frac{0-1}{-(a+j\omega)} = \frac{1}{a+j\omega}</math></p> <p>b) <math>\delta(t)</math>  <math display="block">X(j\omega) = \int_{-\alpha}^{\alpha} \delta(t) e^{-j\omega t} dt = e^{-j\omega 0} = 1</math></p> <p>c) <math>\cos(\omega_0 t)</math>  <math display="block">X(j\omega) = \int_{-\alpha}^{\alpha} \cos(\omega_0 t) e^{-j\omega t} dt</math> <math display="block">= \int_0^{\alpha} \frac{e^{-j\omega_0 t} + e^{j\omega_0 t}}{2j} e^{-j\omega t} dt</math> <math display="block">= \frac{1}{2j} \cdot (F[e^{-j\omega_0 t}] + F[e^{j\omega_0 t}])</math> <math display="block">= \frac{1}{2j} (2\pi \cdot \delta(\omega + \omega_0) + 2\pi \cdot \delta(\omega - \omega_0))</math> <math display="block">-j\pi(\delta(\omega + \omega_0) + \delta(\omega - \omega_0))</math></p>	2	7	
XI	<p>a) <math>\frac{1}{(s-3)(s-2)}</math>            let <math>\frac{A}{s-3} + \frac{B}{s-2} = \frac{A}{s-3} + \frac{B}{s-2}</math> then <math>\frac{A(s-2)+B(s-3)}{s-3} = \frac{A}{s-3} + \frac{B}{s-2}</math>            or <math>A(s-2) + B(s-3) = 1</math>            when <math>s = 2, A(2-2) + B(2-3) = 1 \rightarrow -B = 1</math>            or <math>\rightarrow B = -1</math>            when <math>s = 3, A(3-2) + B(3-3) = 1 \rightarrow A = 1</math>            Now the expression becomes  <math display="block">\frac{1}{s-3} - \frac{1}{s-2}</math>            The inverse Laplace transform is  <math display="block">(e^{3t} - e^{2t})u(t)</math></p> <p>b) <math>\frac{s-a}{(s-a)^2 + b^2}</math>            the Laplace transform of <math>\cos(bt)</math> is  <math display="block">\frac{s}{s^2 + b^2}</math></p>	3	7	
		2		

	<p>Using first shifting property of Laplace transform  Laplace transform of <math>e^{at}\sin(bt)</math> is</p> $\frac{s-a}{(s-a)^2+b^2}$ <p>Inverse Laplace transform is <math>e^{at}\sin(bt)</math></p> <p>c) <math>\frac{1}{s^2+2bs+b^2} = \frac{1}{(s+b)^2}</math>  the Laplace transform of <math>\frac{1}{s^2}</math> is</p> $tu(t)$ <p>using first shift property of Laplace transform the Laplace transform of <math>te^{-at}u(t)</math> is <math>\frac{1}{(s+b)^2}</math>.</p> <p>Therefore, the answer is <math>te^{-at}u(t)</math></p>	2		
XII	Function	Laplace Transform		7
	a) $\cos(bt)$	$\frac{s}{s^2+b^2}$	1	
	b) $e^{-at}u(t)$	$\frac{1}{s+a}$	1	
	c) $\sin(bt)$	$\frac{b}{s^2+b^2}$	1	
	d) $\delta(t)u(t)$	1	1	
	e) $r(t)$	$\frac{1}{s^2}$	1	
	f) $p(t)$	$\frac{2}{s^3}$	1	
	g) $u(t)$	$\frac{1}{s}$	1	
XIII	<p>Time Scaling:</p> $f\left(\frac{t}{a}\right) \xleftrightarrow{\mathcal{L}} aF(as)$ <p>Linearity</p> $f_1(t) \xrightarrow{\mathcal{L.T.}} F_1(s) \text{ with } ROC = R_1$ $f_2(t) \xrightarrow{\mathcal{L.T.}} F_2(s) \text{ with } ROC = R_2$ $af_1(t) + bf_2(t) \xrightarrow{\mathcal{L.T.}} aF_1(s) + bF_2(s): ROC = R_1 \cap R_2$ <p>Time Shifting</p> $f(t) \xrightarrow{\mathcal{L.T.}} F(s) \text{ with } ROC = R$ $f(t-t_0) \xrightarrow{\mathcal{L.T.}} e^{-st_0} F(s): ROC = R$	<p>1</p> <p>+</p> <p>2</p> <p>+</p> <p>2</p> <p>+</p>		7

	<p>Initial Value Theorem</p> <p>Initial value theorem is applied when in Laplace transform the degree of the numerator is less than the degree of the denominator</p> $f(0) = \lim_{s \rightarrow \infty} sF(s)$	2		
XIV	<p>Definition 2 marks illustration and explanation 5 marks</p> <p>Region of Convergence (ROC) is defined as the set of points in s-plane for which the Laplace transform of a function x(t) converges. In other words, the range of Re(s) (i.e. <math>\sigma</math>) for which the function X(s) converges is called the region of convergence.</p> <div style="display: flex; justify-content: space-around; align-items: flex-end;"> <div style="text-align: center;">  <p>Causal</p> </div> <div style="text-align: center;">  <p>Stable</p> </div> <div style="text-align: center;">  <p>Unstable</p> </div> </div>	2+5	7	