

SCHEME OF VALUATION (Scoring Indicators)

Revision : 2015

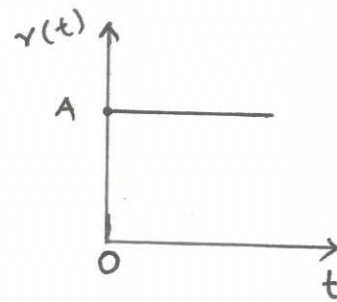
Course code: 5043

Course Title : CONTROL SYSTEMS

Qst No	Scoring Indicator	Split up Score	Sub Total	Total														
<u>PART-A</u>																		
I	1	$\frac{1}{s^2}$	2	2														
	2	$\frac{s+2}{(s+2)^2+9}$	2	2														
	3	Transfer function is the ratio of Laplace transform of output to Laplace transform of input with zero initial condition	2	2														
	4	If the system is stable for all the range of system component values then the system is said to be absolutely stable system.	2	2														
	5	The path taken by a root of characteristic equation when open loop gain K is varied from zero to infinity.	2	2														
<u>PART-B</u>																		
II	1	<table style="width: 100%; border: none;"> <tr> <td style="text-align: center;"><u>Open loop system</u></td> <td style="text-align: center;"><u>closed loop system</u></td> </tr> <tr> <td>a) Inaccurate and unreliable</td> <td>- Accurate and reliable</td> </tr> <tr> <td>b) Feed back is absent</td> <td>- Feed back is present</td> </tr> <tr> <td>c) Stable</td> <td>- Unstable</td> </tr> <tr> <td>d) Simple and economical</td> <td>- Complex and costly</td> </tr> <tr> <td>e) Highly sensitive to disturbances</td> <td>- Less sensitive</td> </tr> <tr> <td>f) Changes in o/p due to external disturbances are not corrected automatically</td> <td>- Changes in output due to external disturbances are corrected automatically</td> </tr> </table>	<u>Open loop system</u>	<u>closed loop system</u>	a) Inaccurate and unreliable	- Accurate and reliable	b) Feed back is absent	- Feed back is present	c) Stable	- Unstable	d) Simple and economical	- Complex and costly	e) Highly sensitive to disturbances	- Less sensitive	f) Changes in o/p due to external disturbances are not corrected automatically	- Changes in output due to external disturbances are corrected automatically	6	6
<u>Open loop system</u>	<u>closed loop system</u>																	
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	a) <u>Step signal</u>	Signal whose values changes from zero to A at $t=0$ and remains constant at A for $t>0$																

$$r(t) = A ; t \geq 0$$

$$= 0 ; t < 0$$

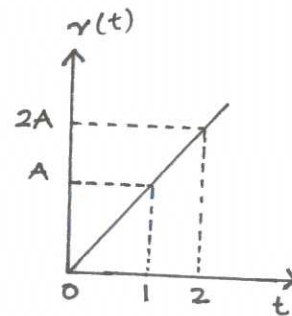


b) Ramp Signal

Signal whose values increases linearly with time from an initial value of zero at $t=0$

$$r(t) = At ; t \geq 0$$

$$= 0 ; t < 0$$



2

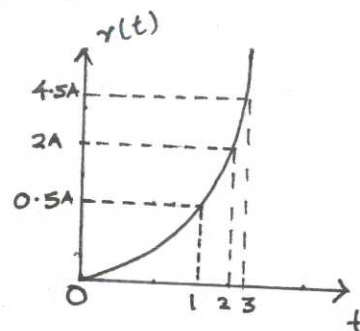
3

c) Parabolic Signal

Signal in which the instantaneous value varies as square of the time from an initial value of zero at $t=0$

$$r(t) = \frac{At^2}{2} ; t \geq 0$$

$$= 0 ; t < 0$$

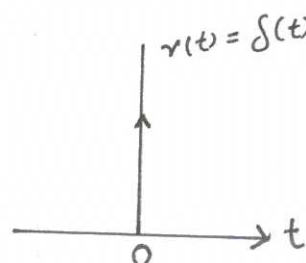


d) Impulse Signal

A signal of very large magnitude which is available for very short duration

$$\delta(t) = \infty ; t = 0$$

$$= 0 ; t \neq 0$$



3

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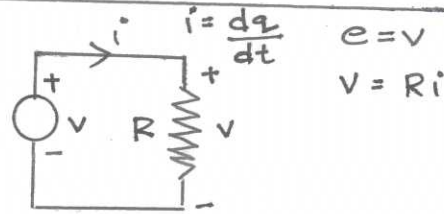
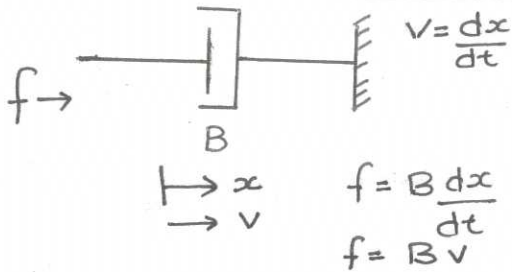
Force - Voltage Analogy

Mechanical system

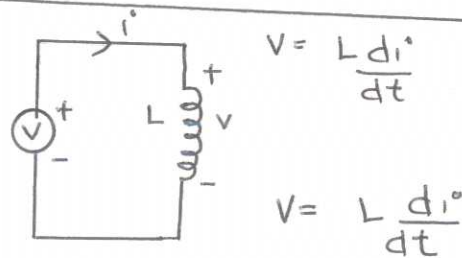
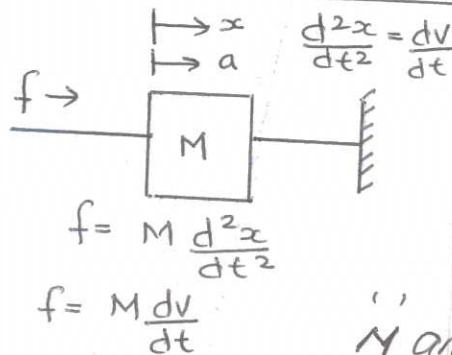
Electrical system

input : Force
output : Velocity

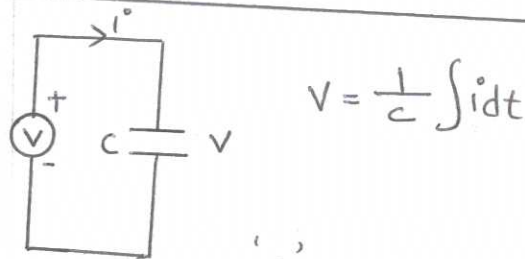
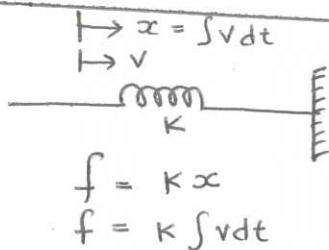
input : Voltage source
output : current



B analogous to R



M analogous to L



K analogous to $\frac{1}{C}$

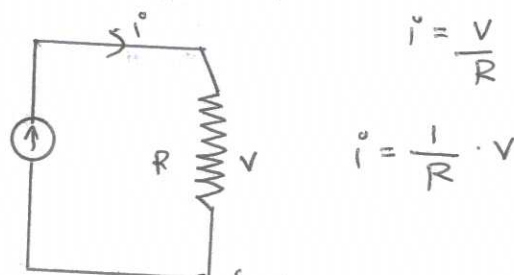
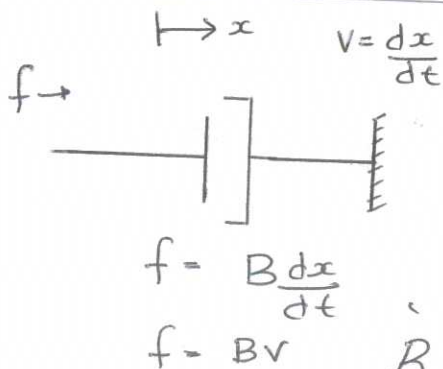
Force - Current Analogy

Mechanical system

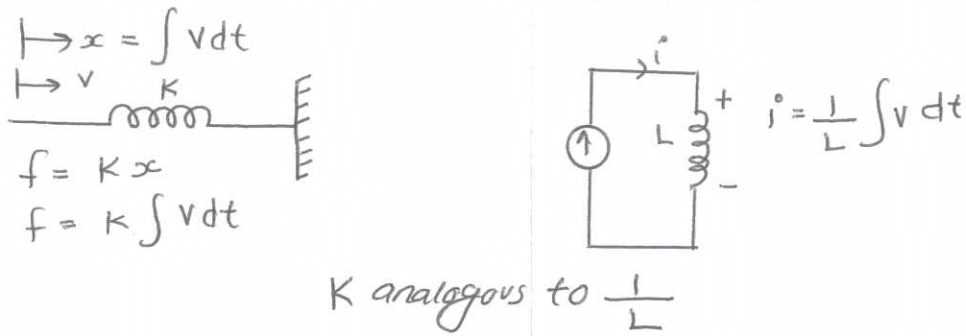
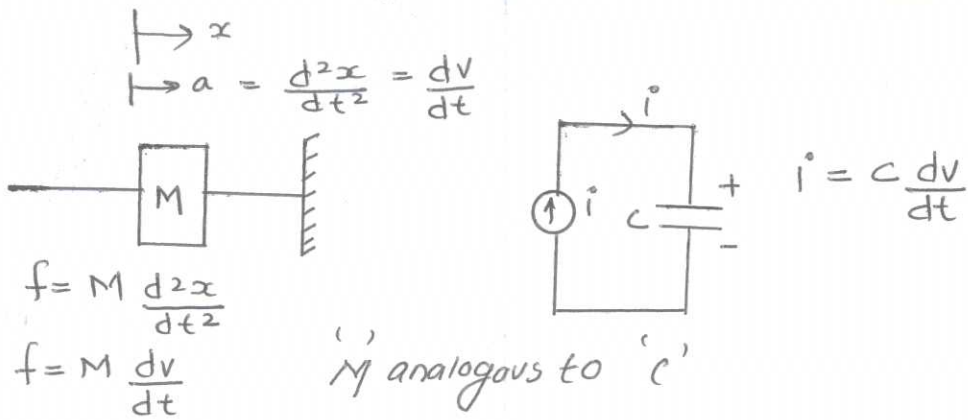
Electrical system

input : Force
output : Velocity

input : current
output : Voltage



B analogous to $\frac{1}{R}$



- 4 (a) Transmittance :- The gain acquired by the signal when it travels from one node to another 2
- (b) Individual loop :- closed path starting from a node and ending at the same node and it does not cross any node more than once 2
- (c) Forward path :- Path starts from input node and ends at output node which does not cross any node more than once 2

5

Characteristic equation $s^4 + 3s^3 + 3s^2 + 2s + 1 = 0$

s^4	1	3	1
s^3	3	2	0
s^2	7/3	1	0
s^1	5/7	0	0
s^0	1	0	0

Location of roots :- All roots are laying on left half of 's' plane 1

The system is stable 1

$$G(s)H(s) = \frac{K}{s(s+2)(s+4)}$$

No. of poles $P = 3$ $P-Z = 3$

No. of Zeros $Z = \text{NIL}$

Poles = $0, -2, -4$

$$\text{Centroid} = \frac{\text{Sum of poles} - \text{Sum of Zeros}}{P-Z}$$

$$= \frac{0 - 2 - 4 - 0}{3} = \frac{-6}{3} = \underline{\underline{-2}} \quad 3$$

6

$$\text{Angle of asymptotes} = \frac{(2q+1)180^\circ}{P-Z} \quad q = 0, 1, 2$$

$$= \frac{(2 \times 0 + 1)180^\circ}{3} = 60^\circ$$

$$\frac{(2 \times 1 + 1)180^\circ}{3} = 180^\circ$$

$$\frac{(2 \times 2 + 1)180^\circ}{3} = 300^\circ \text{ or } -60^\circ \quad 3$$

Centroid = -2

Angle of asymptotes are $60^\circ, 180^\circ, 300^\circ$

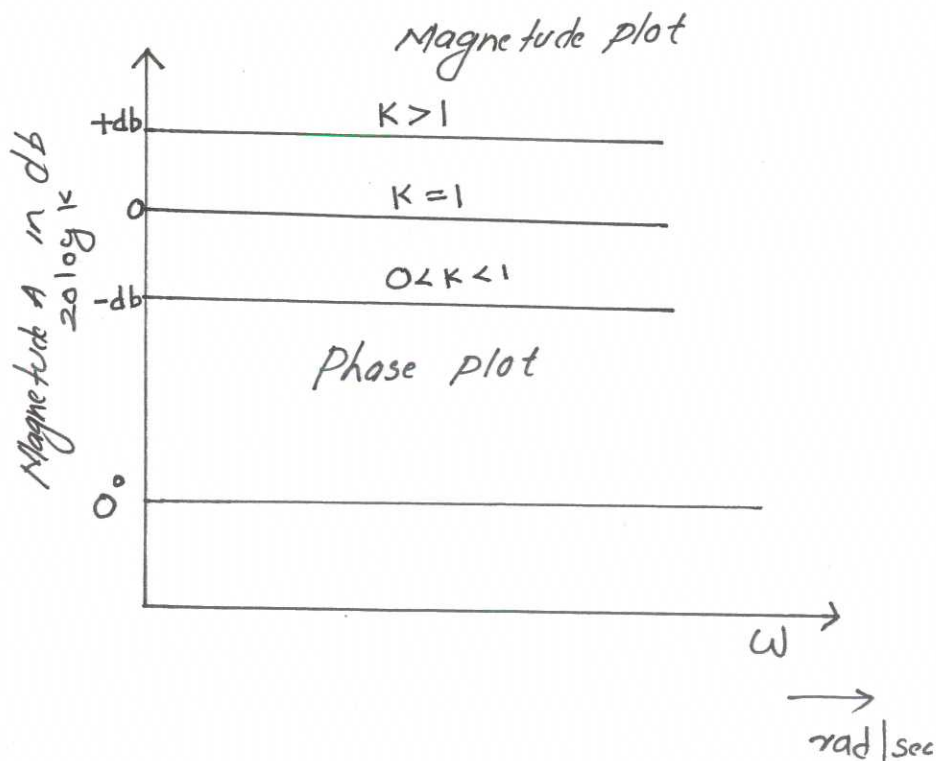
Let $G(s) = K$

$$G(j\omega) = K = K \angle 0^\circ$$

$$\text{Magnitude } A = |G(j\omega)|_{\text{in db}} = 20 \log K$$

$$\text{Phase angle } \phi = \angle G(j\omega) = 0^\circ$$

7



3

30

6

3

PART-C

(i) cos at
 $L \{ f(t) \} = \int_0^{\infty} e^{-st} f(t) dt$
 $L \{ \cos at \} = \int_0^{\infty} e^{-st} \cos at dt$

We have $\int e^{at} \cos bt = \frac{e^{at}}{a^2+b^2} [a \cos bt + b \sin bt]$ 2

By comparing $a = -s, b = a$

$$= \left[\frac{e^{-st}}{s^2+a^2} (-s \cos at + a \sin at) \right]_0^{\infty}$$

$$= \frac{e^{-\infty}}{s^2+a^2} [-s \cos \infty + a \sin \infty] - \frac{e^{-0}}{s^2+a^2} [-s \cos 0 + a \sin 0]$$

$L \{ \cos at \} = \frac{s}{s^2+a^2}$

ii) e^{at}

$$L \{ e^{at} \} = \int_0^{\infty} e^{-st} \cdot e^{at} dt$$

$$= \int_0^{\infty} e^{-t(s-a)} dt$$

$$= \left[\frac{e^{-t(s-a)}}{-(s-a)} \right]_0^{\infty} = \frac{1}{s-a}$$

iii) 't'

$$L \{ t \} = \int_0^{\infty} e^{-st} \cdot t dt$$

$$= t \cdot \int_0^{\infty} e^{-st} - \int_0^{\infty} \left[\frac{d}{dt} (t) \cdot \int_0^{\infty} e^{-st} \right] \text{ by product rule}$$

$$= t \cdot \frac{e^{-st}}{-s} - \int_0^{\infty} 1 \cdot \frac{e^{-st}}{-s}$$

$$= \left[\frac{t e^{-st}}{-s} - \frac{1}{s^2} e^{-st} \right] - \left[0 \cdot \frac{e^{-0}}{s} - \frac{1}{s^2} e^{-0} \right]$$

$$= \frac{1}{s^2}$$

III a

b

$$(b) \quad \frac{2s^2 - 4}{(s+1)(s-2)(s-3)}$$

$$\text{Let } \frac{2s^2 - 4}{(s+1)(s-2)(s-3)} = \frac{A}{s+1} + \frac{B}{s-2} + \frac{C}{s-3}$$

$$2s^2 - 4 = A(s-2)(s-3) + B(s+1)(s-3) + C(s+1)(s-2)$$

$$\text{put } s = -1$$

$$A = \frac{-1}{6}$$

$$\text{put } s = 2$$

$$B = \frac{-4}{3}$$

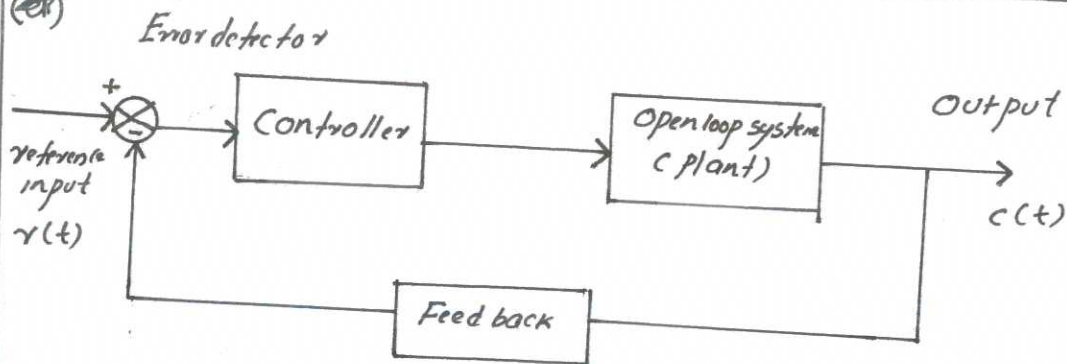
$$\text{put } s = 3$$

$$C = \frac{7}{2}$$

$$\frac{2s^2 - 4}{(s+1)(s-2)(s-3)} = \frac{-1}{6(s+1)} - \frac{4}{3(s-2)} + \frac{7}{2(s-3)}$$

$$= \frac{-1}{6} e^{-t} - \frac{4}{3} e^{\frac{2t}{3}} + \frac{7}{2} e^{\frac{3t}{2}}$$

(ex)



- Control system in which the output has an effect up on input quantity in order to maintain the desired output value are called closed loop system
- The feed back automatically corrects the change in output due to disturbances. Hence closed loop system also called Automatic control system
- closed loop system consist of an error detector, a controller, plant and feed back element
- The reference signal (i.e. signal) corresponds to desired output.
- The feed back path elements samples the output and convert it to a signal of same type as that of reference signal

The error signal generated by error detector is the difference between reference signal and feedback signal.

- Controller modifies and amplifies the error signal
- Modified error signal is fed to the plant to correct its output.

IV b

(b)

$$L\{y\} = Y(s)$$

$$L\{y'\} = sY(s) - y(0)$$

$$L\{y''\} = s^2Y(s) - y(0) - y'(0)$$

$$L\{y'''\} = s^3Y(s) - y(0) - y'(0) - y''(0)$$

$$L\{e^{2t}\} = \frac{1}{s-2}$$

Given $y'' - y = e^{2t}$

$$L\{y'' - y\} = L\{e^{2t}\}$$

$$[s^2Y(s) - y(0) - y'(0)] - Y(s) = \frac{1}{s-2}$$

$$Y(s) = \frac{1}{(s-2)(s+1)}$$

By partial fraction

$$\frac{1}{(s-2)(s+1)} = \frac{A}{s-2} + \frac{B}{s+1}$$

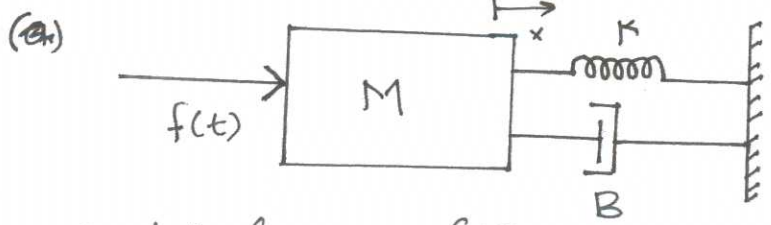
$$A = -1$$

$$B = -1/3$$

$$\frac{1}{(s-2)(s+1)} = \frac{1/3}{s-2} + \frac{-1/3}{s+1}$$

$$= \frac{1}{3}e^{2t} - \frac{1}{3}e^{-t}$$

V a



Applied force = $f(t)$

Total opposing force = $f_m + f_b + f_k$

2

9

3

4

2

15

1

$$f_m = M \frac{d^2x}{dt^2}, F_b = B \frac{dx}{dt}, f_k = kx$$

Newton's 2nd law $f(t) = f_m + f_b + f_k$

$$= M \frac{d^2x}{dt^2} + B \frac{dx}{dt} + kx$$

Taking L.T both Side

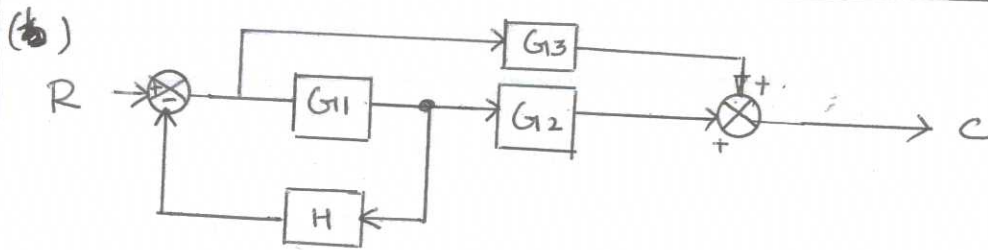
$$F(s) = X(s) [Ms^2 + Bs + K]$$

$$T.F = \frac{X(s)}{F(s)} = \frac{1}{Ms^2 + Bs + K}$$

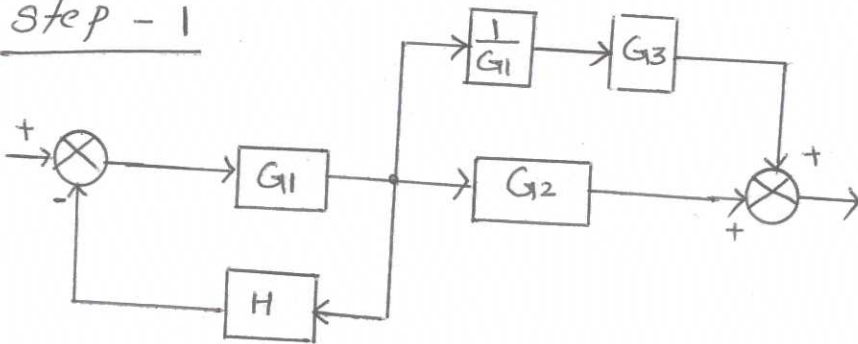
1

6

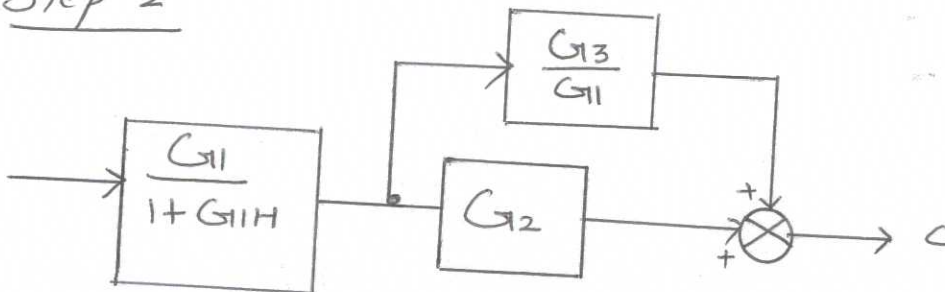
2



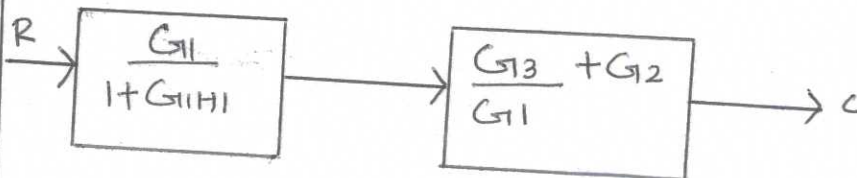
Step - 1



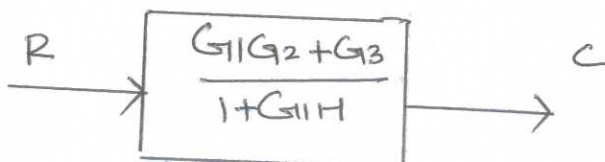
Step 2



Step - 3



Step - 4



$$\frac{C}{R} = \frac{G_{11}G_2 + G_{13}}{1 + G_{11}H}$$

2

2

9

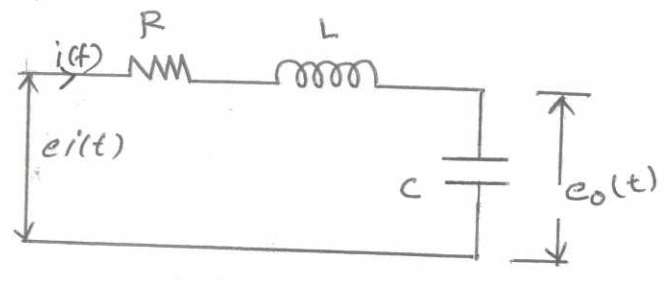
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3

V b

VI

(a)



(a)

Here $V_2 = e_o(t)$

Apply KVL $e^{i\omega t} = i(t)R + L \frac{di(t)}{dt} + \frac{1}{C} \int i(t) dt$

$$E i(s) = I(s)R + Ls I(s) + \frac{1}{Cs} I(s)$$

$$E i(s) = I(s) \left[R + Ls + \frac{1}{Cs} \right] \quad \text{--- (1)}$$

$$e_o(t) = \frac{1}{C} \int i(t) dt$$

$$E_o(s) = \frac{1}{Cs} I(s)$$

$$I(s) = \frac{E i(s)}{R + Ls + \frac{1}{Cs}}$$

$$E_o(s) = \frac{E i(s)}{Rcs + Lcs^2 + 1}$$

$$TF = \frac{1}{Rcs + Lcs^2 + 1}$$

1

6

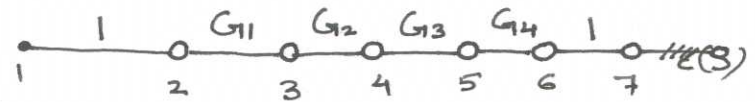
4

1

VI

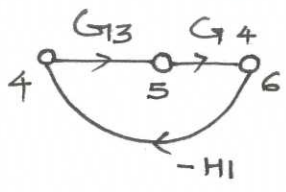
(b) Forward path gain

Only one forward path $K=1$

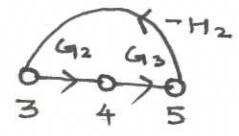


Gain = $G_{11} G_{12} G_{13} G_{14}$

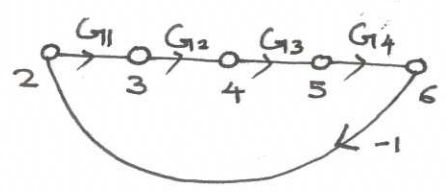
Individual loop Gain



$P_{11} = -G_{13} G_{14} H_1$



$P_{21} = -G_{12} G_{13} H_2$



$P_{31} = -G_{11} G_{12} G_{13} G_{14}$

- No possible combination of two non touching loops
three non touching loops etc

$$\Delta = 1 - (P_{11} + P_{21} + P_{31})$$

$$= 1 + G_{13} G_{14} H_1 + G_{12} G_{13} H_2 + G_{11} G_{12} G_{13} G_{14}$$

$\Delta_1 = 1$

1

1

1

1

1

15

Transfer function T

$$T = \frac{C(s)}{R(s)} = \frac{1}{\Delta} \sum_k P_k \Delta_k$$

$$= \frac{1}{\Delta} P_1 \Delta_1 \quad \therefore k=1$$

$$= \frac{G_1 G_2 G_3 G_4}{1 + G_3 G_4 H_1 + G_2 G_3 H_2 + G_1 G_2 G_3 G_4}$$

2

9

2

VII

a

(*) $H(s) = 1$

$$K_p = \lim_{s \rightarrow 0} G(s) H(s) = \lim_{s \rightarrow 0} \frac{10(s+2)}{s^2(s+1)} \cdot 1 = \frac{20}{0} = \infty$$

$$K_v = \lim_{s \rightarrow 0} s G(s) H(s) = \lim_{s \rightarrow 0} s \frac{10(s+2)}{s^2(s+1)} = \frac{20}{0} = \infty$$

$$K_a = \lim_{s \rightarrow 0} s^2 G(s) H(s) = \lim_{s \rightarrow 0} \frac{10(s+2)}{s+1} = \frac{20}{1} = 20$$

$$K_a = \frac{10 \times 2}{1} = \underline{\underline{20}}$$

2

2

6

2

VII

b

(*)

s^6	1	8	20	16
s^5	1	6	8	0
s^4	1	6	8	
s^3	0	0		
s^2	1	3		
s^1	3	8		
s^0	0.33	0		
s^0	8			

Auxiliary equation

$$A = s^4 + 6s^2 + 8$$

$$\frac{dA}{ds} = 4s^3 + 12s$$

- No sign change

- system is marginally stable

5

9

2

1

1

15

out

$$r(t) = t, \quad R(s) = \frac{1}{s^2}$$

$$C(s) = \frac{1}{1+ST} R(s)$$

$$\text{put } R(s) = \frac{1}{s^2}$$

$$C(s) = \frac{1}{s^2(1+ST)}$$

By partial fraction

$$C(s) = \frac{1}{s^2(1+ST)} = \frac{As+B}{s^2} + \frac{C}{ST+1}$$

$$A = -T$$

$$B = 1$$

$$C = T^2$$

$$\therefore \frac{-Ts+1}{s^2} + \frac{T}{ST+1}$$

$$L^{-1} \left\{ \frac{-Ts+1}{s^2} + \frac{T}{ST+1} \right\}$$

$$= -T + t + Te^{-t/T}$$

Steady State error :- It is the value of $e(t)$ when t tends to infinity

$$\therefore e_{ss} = \lim_{t \rightarrow \infty} e(t)$$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s R(s)}{1+G(s)H(s)}$$

For unit step input $R(s) = \frac{1}{s}$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s \times 1/s}{1+G(s)H(s)} = \frac{1}{1+K_p}$$

$K_p \rightarrow$ Positional error constant

Type-0 system

$$K_p = \lim_{s \rightarrow 0} G(s)H(s) = \lim_{s \rightarrow 0} K \frac{(s+z_1)(s+z_2)(s+z_3)}{(s+p_1)(s+p_2)(s+p_3)}$$

$$= K \frac{z_1 z_2 z_3}{p_1 p_2 p_3} = \text{Constant}$$

$$e_{ss} = \frac{1}{1+K_p} = \text{Constant}$$

Type - 1 system

$$K_p = \lim_{s \rightarrow 0} G(s)H(s) = \lim_{s \rightarrow 0} \frac{K(s+z_1)(s+z_2)(s+z_3)}{s(s+p_1)(s+p_2)(s+p_3)} \quad 2$$

$$e_{ss} = \frac{1}{1+K_p} = \frac{1}{1+\infty} = 0$$

Type 2 system

$$K_p = \lim_{s \rightarrow 0} G(s)H(s) = \lim_{s \rightarrow 0} \frac{K \cdot (s+z_1)(s+z_2)(s+z_3)}{s^2(s+p_1)(s+p_2)(s+p_3)} \quad 2$$

$$e_{ss} = \frac{1}{1+K_p} = \frac{1}{1+\infty} = 0$$

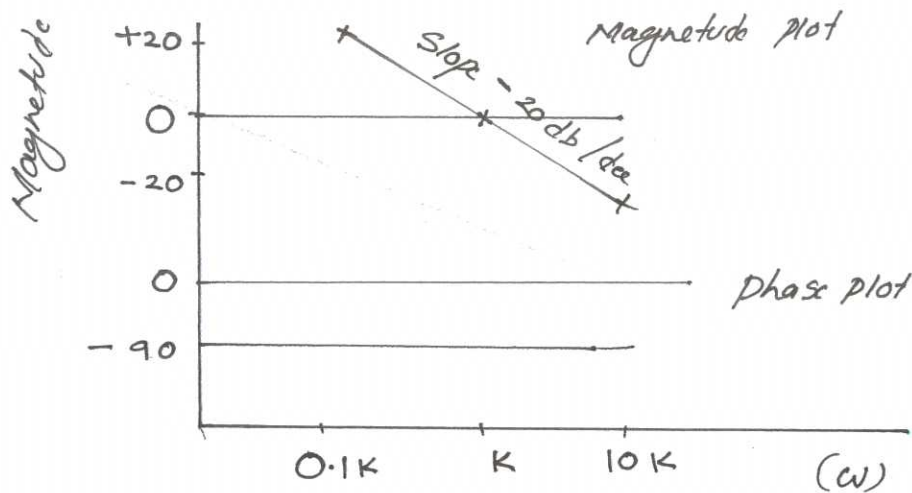
Magnitude $A = |G(j\omega)| = 20 \log \left(\frac{K}{\omega} \right)$

$\phi = \angle G(j\omega) = -90^\circ$

When $\omega = 0.1K$, $A = 20 \log \left(\frac{1}{0.1} \right) = 20 \text{ db}$

$\omega = K$ $A = 20 \log \frac{K}{K} = 0 \text{ db}$ 3

$\omega = 10K$ $A = 20 \log \frac{K}{10K} = -20 \text{ db}$



Step 1 - : Locate the poles and zeros of $G(s)H(s)$ on the s plane

Step 2 - : Determine root locus on real axis

Step 3 - : Determine Asymptotes of root locus and meeting point with real axis

Step 4 - : Find break away and break in points

Step 5 - : Find the angle of departure and angle of arrival

Step 6 - : Find the crossing point of root locus crosses the imaginary axis

X

b

(i) Gain Margin - : The value by which the gain of the system has to be increased to drive system to be verge of instability.

- It is given by the reciprocal of the magnitude of open loop transfer function at phase cross over frequency

$$K_g = \frac{1}{|G(j\omega)|_{\omega=\omega_{pc}}} \quad \text{and}$$

$$K_g \text{ in db} = 20 \log \frac{1}{|G(j\omega)|_{\omega=\omega_{pc}}} \\ = -20 \log |G(j\omega)|_{\omega=\omega_{pc}}$$

(ii)

Phase Margin - : Amount of additional phase lag at ~~given~~ gain cross over frequency ω_{gc} required to bring the system to the verge of instability.

$$It \text{ is } 180 + \phi_{gc}$$

(iii)

Phase Cross Over frequency - : Frequency at which the phase of the open loop transfer function is 180°

(iv)

Gain cross Over frequency - : Frequency at which magnitude of the open loop transfer function is unity

5

2

9

2